## Mark schemes

1. (a) difference amplifier $\checkmark$
(b) $\quad V_{\text {out }}=\left(V_{+}-V_{-}\right) \times\left(R_{f} / R_{\text {in }}\right)$

$$
\begin{aligned}
& \mathrm{V}_{\text {out }}=(0 \mathrm{~V}-150 \mathrm{mV}) \times(1 \mathrm{M} \Omega / 100 \mathrm{k} \Omega) \checkmark \\
& \mathrm{V}_{\text {out }}=-1.5 \mathrm{~V} \checkmark
\end{aligned}
$$

1 mark for the correct resistor substitution / resistor ratio (10)
1 mark for -1.5 V (must have correct sign)
(c) Signal 2 is subtracted from signal 1 by the difference amplifier $\checkmark$

Noise is common to both so will be reduced / eliminated when subtracted $\checkmark$
Signals will also be subtracted resulting in an addition (re-enforcement) of the signal. $\checkmark$

## Accept arguments based on the 'phase' relationship

2. 

(a)

(b) The non-inverting input
(non-inverting)
(c) $\quad \mathrm{I}=\left(\mathrm{V}_{\text {in }}-\mathrm{V}_{\mathrm{x}}\right) / \mathrm{R}_{\text {in }}=\left(\mathrm{V}_{\mathrm{x}}-\mathrm{V}_{\text {out }}\right) / \mathrm{R}_{\mathrm{f}}$

But $\mathrm{V}_{\mathrm{x}}=0 \mathrm{~V}$ (a virtual earth)
$I=V_{\text {in }} / R_{\text {in }}=-V_{\text {out }} / R_{f}$
Making use of: $I_{\text {in }}=-I_{F}$
$\frac{V_{\text {out }}}{V_{\text {in }}}=\frac{R_{f}}{R_{\text {in }}}$
(d) Voltage gain (Channel 1) $=-R_{f} / R_{\text {in }} 1$
$-(150 \mathrm{k} \Omega / 7.5 \mathrm{k} \Omega)$
$-20$
Both number and sign must be correct
(e) $\quad V_{\text {out }}=-R_{f}\left(V_{\text {in Ch1 }} / R_{1}+V_{\text {in Ch2 }} / R_{2}\right)$
$=-150 \mathrm{k} \Omega((15 \mathrm{mV} / 7.5 \mathrm{k} \Omega)+(-100 \mathrm{mV} / 30 \mathrm{k} \Omega))$
$=-((0.3)+(-0.5))=0.2$ Volts
Evidence of correct method
Answer and correct sign
(f) By using variable resistors

The gain can easily be changed
or
the relative levels of the two channels can be set
or
the required balance between the two signals can be made
One relevant point made
3. (a) Voltage in = Voltage out / Voltage gain
$=3 \mathrm{~V} / 40$
$=75 \times 10^{-3} \mathrm{~V}$ V
(b)


Two resistor chain, correctly labelled connected between output and ground
Inverting input connected to mid-point of resistor chain
(c) $\frac{V_{\text {out }}}{V_{\text {in }}}=1+\frac{R_{f}}{R_{\text {in }}}$
$40=1+\frac{R_{f}}{R_{\text {in }}}$ calculation to give resistor ratio of $39 \checkmark$
$R_{\text {in }}=1 \mathrm{k} \Omega ; R_{\mathrm{f}}=39 \mathrm{k} \Omega \checkmark$
(d) Desired gain $\times$ bandwidth is $40 \times 50 \mathrm{kHz}=2 \mathrm{MHz} \checkmark$

The Op Amp can only provide $1 / 2$ the amplification needed. Not suitable. $\checkmark$
1 mark - relevant calculation
1 mark - reference to only providing $1 / 2$ require amplification / gain so not suitable
4. (a) Photoconductive mode Accept 'reverse bias'
(b) Dark currents will become a source of noise - need to keep $\mathrm{S}: \mathrm{N}$ as high as possible OWTTE

OR
Need to have a large difference in signal when detector is in light and dark $\checkmark$ Must include idea of 'noise'
OR
Must include concept of large signal change to represent digital signal
(c) At $850 \mathrm{~nm}, R_{\lambda}=0.50 \mathrm{~A} / \mathrm{W} \checkmark$

Reading from graph
Allow 0.49 A/W to 0.51 A/W
Using $R_{\lambda}=\frac{I_{\mathrm{p}}}{P} \quad I_{\mathrm{p}}=R_{\lambda} \times P \quad 0.50 \times 4 \times 10^{-6}=2 \mu \mathrm{~A} \checkmark$ ecf
$V_{\text {out }}=I_{\mathrm{p}} \times R 2 \mu \mathrm{~A} \times 560 \mathrm{k} \Omega=+1.12 \mathrm{~V} \checkmark$
Accept voltage in range of 1.10 V to 1.14 V
Accept value without + sign
(d)


Correct configuration of $R_{1}$ and $R_{2} \checkmark$
$R_{1}: R_{2}$ ratio $3: 1$ in suggested range $\checkmark$
Label the input point which must have a direct connection to the non-inverting input $\sqrt{ }$
One mark only
An inverting op amp configuration with a voltage gain -4.

